Randomized Time Division strategy for broadcasting information

Chandra Nair

Department of Information Engineering

The Chinese University of Hong Kong chandra.nair@gmail.com

Feb 26, 2014

BROADCAST CHANNELS [COVER '72]

Figure : Discrete memoryless broadcast channel

- How to simultaneously communicate independent messages to the two receivers over a shared medium?
- Want to take advantage of the fact that the transmitted symbols are corrupted differently for the two receivers

ENCODING STRATEGY [COVER '72]

A natural coding strategy is TIME DIVISION. Can one do better?

- Johnny English understands only English alphabets: A, B and when it is
- Ms. Discrete understands understands only 0 and 1, and when it is

AABBAABB101110 or 101110AABBAABB

ENCODING STRATEGY [COVER '72]

A natural coding strategy is TIME DIVISION.

Can one do better?

Imagine that you are simultaneously speaking to two people:

- Johnny English understands only English alphabets: A, B and when it is not either he denotes it as ∗.
- Ms. Discrete understands understands only 0 and 1, and when it is neither she denotes it as ?.

Suppose you want to communicate a word AABBAABB to Johnny English and the word 101110 to Ms. Discrete then your can either speak

AABBAABB101110 or 101110AABBAABB

ENCODING STRATEGY [COVER '72]

A natural coding strategy is TIME DIVISION.

Can one do better?

Imagine that you are simultaneously speaking to two people:

- Johnny English understands only English alphabets: A, B and when it is not either he denotes it as ∗.
- Ms. Discrete understands understands only 0 and 1, and when it is neither she denotes it as ?.

Suppose you want to communicate a word AABBAABB to Johnny English and the word 101110 to Ms. Discrete then your can either speak

AABBAABB101110 or 101110AABBAABB

In particular you can choose any one of the $\binom{14}{8}$ $\binom{14}{8}$ = 3003 possibilities Which particular order you choose can convey $\lfloor log_2(3003)\rfloor = 11$ additional bits of information that is decodable by both receivers. Such a strategy is called RANDOMIZED TIME DIVISION

COMPARING ACHIEVABLE RATE REGIONS

Using time-division one can achieve

$$
R_1 \leq \alpha C_{JE} = \alpha \log_2(3)
$$

\n
$$
T_2 \leq (1 - \alpha)C_D = (1 - \alpha) \log_2(3)
$$

for any $0 \leq \alpha \leq 1$.

Using randomized time-division, one can achieve (R_1, R_2) satisfying

$$
R_1 \le H(\alpha) + \alpha \log_2(2)
$$

\n
$$
R_2 \le H(\alpha) + (1 - \alpha) \log_2(2)
$$

\n
$$
R_1 + R_2 \le H(\alpha) + \alpha \log_2(2) + (1 - \alpha) \log_2(2)
$$

for any $0 \leq \alpha \leq 1$.

RATE REGIONS

Figure : Achievable regions for time division vs randomized time division

RECAP: RANDOMIZED TIME-DIVISION STRATEGY

- One picks a codebook of binary codewords that is decodable by both receivers
- The codeword is picked from this codebook that reveals *common (timing) information* to both receivers
- On the locations where this *timing* codeword has 0, the encoder sends a codeword tailored to be decoded by receiver *Y*
- On the locations where this *timing* codeword has 1, the encoder sends a codeword tailored to be decoded by receiver *Z*

RECAP: RANDOMIZED TIME-DIVISION STRATEGY

- One picks a codebook of binary codewords that is decodable by both receivers
- The codeword is picked from this codebook that reveals *common (timing) information* to both receivers
- On the locations where this *timing* codeword has 0, the encoder sends a codeword tailored to be decoded by receiver *Y*
- On the locations where this *timing* codeword has 1, the encoder sends a codeword tailored to be decoded by receiver *Z*

Using randomized time-division, one can achieve (R_1, R_2) satisfying

 $R_1 \leq I(W;Y) + P(W=0)I(X;Y|W=0)$ $R_2 \leq I(W; Z) + P(W = 1)I(X; Z|W = 1)$ $R_1 + R_2 \le \min\{I(W;Y), I(W;Z)\} + P(W=0)I(X;Y|W=0)$ $+ P(W = 1)I(X; Z|W = 1)$

HOW GOOD IS THIS STRATEGY?

Randomized time divison is a naive improvement over time-division.

The best achievable region for broadcast channel is due to Marton [Mar '79]

Marton's achievable region

The union of rate pairs (R_1, R_2) such that the rate pairs satisfy

 $R_1 \leq I(UW;Y)$ $R_2 \leq I(VW;Z)$ $R_1 + R_2 \leq \min\{I(W; Y), I(W; Z)\} + I(U; Y|W)$ $+ I(V; Z | W) - I(U; V | W)$

for any $(U, V, W) - X - (Y, Z)$ is achievable.

Issue: The above region was not computable (no cardinality bounds on *U*, *V*, *W*)

Conjectured that for a particular binary input broadcast channel

 $SR_{Marton} = SR_{RTD}$ [Nair-Wang '08]

Established cardinality bounds

- Novel kind of perturbation arguments
- Extended the perturbation arguments to show that the earlier conjectured
- Showed that for all *binary* input broadcast channels

Conjectured that for a particular binary input broadcast channel

 $SR_{Marton} = SR_{RTD}$ [Nair-Wang '08]

• Established cardinality bounds

 $|U| \leq |X|, |V| \leq |X|, |W| \leq |X| + 4$

for evaluating Marton's achievable region [Gohari-Anantharam '09]

- Novel kind of perturbation arguments
- Extended the perturbation arguments to show that the earlier conjectured
- Showed that for all *binary* input broadcast channels

Conjectured that for a particular binary input broadcast channel

 $SR_{Marton} = SR_{RTD}$ [Nair-Wang '08]

• Established cardinality bounds

 $|U| < |X|, |V| < |X|, |W| < |X| + 4$

for evaluating Marton's achievable region [Gohari-Anantharam '09]

- Novel kind of perturbation arguments
- Extended the perturbation arguments to show that the earlier conjectured result is true [Jog-Nair '09]
- Showed that for all *binary* input broadcast channels

Conjectured that for a particular binary input broadcast channel

 $SR_{Marton} = SR_{RTD}$ [Nair-Wang '08]

• Established cardinality bounds

 $|U| < |X|, |V| < |X|, |W| < |X| + 4$

for evaluating Marton's achievable region [Gohari-Anantharam '09]

- Novel kind of perturbation arguments
- Extended the perturbation arguments to show that the earlier conjectured result is true [Jog-Nair '09]
- Showed that for all *binary* input broadcast channels

 $SR_{Marton} = SR_{RTD}$ [Geng-Nair-Wang '10]

Conjectured that for a particular binary input broadcast channel

 $SR_{Marton} = SR_{RTD}$ [Nair-Wang '08]

• Established cardinality bounds

 $|U| < |X|, |V| < |X|, |W| < |X| + 4$

for evaluating Marton's achievable region [Gohari-Anantharam '09]

- Novel kind of perturbation arguments
- Extended the perturbation arguments to show that the earlier conjectured result is true [Jog-Nair '09]
- Showed that for all *binary* input broadcast channels

 $SR_{Marton} = SR_{RTD}$ [Geng-Nair-Wang '10]

What about entire region and not just sum-rate?

IMPROVED CARDINALITY BOUNDS

Combining

- perturbation arguments [Gohari-Anantharam]
- concave envelope interpretation for extremal auxiliaries [Nair]

one can show that it suffices to consider auxiliairies *U*, *V* that satisfy

 $|U| + |V| \leq |X| + 1$

to evaluate Marton's achievable region [Anantharam-Gohari-Nair '12]

Marton's achievable region *matches* randomized time division when |*X*|

IMPROVED CARDINALITY BOUNDS

Combining

- perturbation arguments [Gohari-Anantharam]
- concave envelope interpretation for extremal auxiliaries [Nair]

one can show that it suffices to consider auxiliairies *U*, *V* that satisfy

 $|U| + |V| \leq |X| + 1$

to evaluate Marton's achievable region [Anantharam-Gohari-Nair '12]

This shows that when $|X| = 2$, $|U| + |V| < 3$. Since $|U|, |V| \ge 1$; the only possibilities are $|U| = 2, |V| = 1$ or $|V| = 2, |U| = 1.$

This in particular implies that

• Marton's achievable region *matches* randomized time division when |*X*| is binary

MAIN RESULT

 $T_{q,\alpha}(X) := \qquad \text{sup}$ (U,V) → X → (Y,Z) $\alpha I(U;Y) + I(V;Z) - I(U;V)$

Cardinality bounds on *U*, *V*

For all broadcast channels $q(y, z|x)$, for all $\lambda \in [0, 1]$, for all $\alpha \geq 1$, to compute

$$
\mathbf{C}[-(\alpha - \lambda)H(Y) - \lambda H(Z) + T_{q,\alpha}(X)]
$$

it suffices to consider $||U|| + ||V|| \le ||X|| + 1$.

Ⅰ Suffices to consider $||U|| + ||V|| < ||X|| + 1$, $||W|| < ||X|| + 4$ to compute

MAIN RESULT

$$
T_{q,\alpha}(X) := \sup_{(U,V)\to X\to(Y,Z)} \alpha I(U;Y) + I(V;Z) - I(U;V)
$$

Cardinality bounds on *U*, *V*

For all broadcast channels $q(y, z|x)$, for all $\lambda \in [0, 1]$, for all $\alpha \geq 1$, to compute

$$
\mathbf{C}[-(\alpha - \lambda)H(Y) - \lambda H(Z) + T_{q,\alpha}(X)]
$$

it suffices to consider $||U|| + ||V|| \le ||X|| + 1$.

Corollary

1 Suffices to consider $||U|| + ||V|| \le ||X|| + 1$, $||W|| \le ||X|| + 4$ to compute $MIB(q)$

IDEA OF PROOF

Suppose $p(u, v, x)$ is an *extremal distribution* such that $C[-(\alpha - \lambda)H(Y) - \lambda H(Z) + T_{a,\alpha}(X)]$ $= -(\alpha - \lambda)H(Y) - \lambda H(Z) + \alpha I(U;Y) + I(V;Z) - I(U;V),$

then the r.h.s. is *locally concave* with respect to all perturbations of $p(u, v, x)$. Rearrange the right hand side as

 $\lambda(H(Y) - H(Z)) - \alpha H(Y|U) + H(V|U) - H(Z|V)$

$$
p_{\epsilon}(u, v, x) = p(u, v, x)(1 + \epsilon f(u)),
$$

$$
\big(\sum_{u} p(u)f(u) = 0\big).
$$

$$
\frac{d^2}{d\epsilon^2}\left[H(Y)-H(Z)\right]_{\epsilon=0}\leq 0
$$

IDEA OF PROOF

Suppose $p(u, v, x)$ is an *extremal distribution* such that $C[-(\alpha - \lambda)H(Y) - \lambda H(Z) + T_{a,\alpha}(X)]$ $= -(\alpha - \lambda)H(Y) - \lambda H(Z) + \alpha I(U;Y) + I(V;Z) - I(U;V),$

then the r.h.s. is *locally concave* with respect to all perturbations of $p(u, v, x)$. Rearrange the right hand side as

$$
\lambda(H(Y) - H(Z)) - \alpha H(Y|U) + H(V|U) - H(Z|V)
$$

Consider a perturbation of the form

$$
p_{\epsilon}(u, v, x) = p(u, v, x)(1 + \epsilon f(u)), \qquad \left(\sum_{u} p(u)f(u) = 0\right).
$$

For the second derivative to be negative, it must be that the second derivative of the term

$$
\frac{d^2}{d\epsilon^2}\left[H(Y)-H(Z)\right]_{\epsilon=0}\leq 0
$$

IDEA OF PROOF (CNTD...)

Alternately, rearrange the right hand side as

 $(1 - \lambda)(H(Z) - H(Y)) - H(Z|V) + H(U|V) - H(U|Y) - (\alpha - 1)H(Y|U)$

Consider a perturbation of the form

$$
\hat{p}_{\epsilon}(u,v,x) = p(u,v,x)(1+\epsilon g(v)), \qquad \big(\sum_{v} p(v)g(v) = 0\big).
$$

For the second derivative to be negative, it must be that the second derivative of the term

$$
\frac{d^2}{d\epsilon^2} \left[H(Z) - H(Y) \right]_{\epsilon=0} \le 0
$$

OBSERVATION

For a fixed channel $q(y, z|x)$ the term $H(Y) - H(Z)$ depends only on $p(x)$.

Hence, if there exists $f(u)$ and $g(v)$ such that $p_{\epsilon}(x) = \hat{p}_{\epsilon}(x)$ for all $x \in \mathcal{X}$, then one would need to have

$$
\frac{d^2}{d\epsilon^2} [H(Y) - H(Z)]_{\epsilon=0} = 0.
$$

OBSERVATION

For a fixed channel $q(y, z|x)$ the term $H(Y) - H(Z)$ depends only on $p(x)$.

Hence, if there exists $f(u)$ and $g(v)$ such that $p_{\epsilon}(x) = \hat{p}_{\epsilon}(x)$ for all $x \in \mathcal{X}$, then one would need to have

$$
\frac{d^2}{d\epsilon^2} [H(Y) - H(Z)]_{\epsilon=0} = 0.
$$

This will in turn force the convex terms to have zero second derivative as well.

As a consequence, it will turn out that the expression

 $-(\alpha - \lambda)H(Y) - \lambda H(Z) + \alpha I(U;Y) + I(V;Z) - I(U;V)$

will remain unchanged by either of these perturbations.

OBSERVATION

For a fixed channel $q(y, z|x)$ the term $H(Y) - H(Z)$ depends only on $p(x)$.

Hence, if there exists $f(u)$ and $g(v)$ such that $p_{\epsilon}(x) = \hat{p}_{\epsilon}(x)$ for all $x \in \mathcal{X}$, then one would need to have

$$
\frac{d^2}{d\epsilon^2} [H(Y) - H(Z)]_{\epsilon=0} = 0.
$$

This will in turn force the convex terms to have zero second derivative as well.

As a consequence, it will turn out that the expression

 $-(\alpha - \lambda)H(Y) - \lambda H(Z) + \alpha I(U;Y) + I(V;Z) - I(U;V)$

will remain unchanged by either of these perturbations.

Set ϵ large enough so that the support of *U* or *V* reduces by one.

CONDITIONS FOR EXISTENCE OF $f(u)$, $g(v)$

- $\sum_{u,v} p(u,v,x) f(u) = \sum_{u,v} p(u,v,x) g(v) \quad \forall x \in \mathcal{X}.$
	- From the condition: $p_{\epsilon}(x) = \hat{p}_{\epsilon}(x)$ for all $x \in \mathcal{X}$.
- 2 $\sum_{u,v,x} p(u,v,x) f(u) = 0.$
	- From the condition: $p_e(x)$ is a valid probability distribution.
- $\sum_{u,v,x} p(u,v,x)g(v) = 0.$
	- From the condition: $\hat{p}_{\epsilon}(x)$ is a valid probability distribution.

So there are $||X|| + 1$ linear constraints on a vector of size $||U|| + ||V||$.

A non-trivial solution exists when $||U|| + ||V|| > ||X|| + 1$.

COMMENTS

It is reasonably straightforward to map this bound on auxiliaries to Marton's inner bound.

The bound $\|W\| \le \|X\| + 4$ comes from convexification (traditional Caratheodory based arguments)

This bound seems to indicate a trade-off between communicating with one receiver vs. the other

• In some sense, the essence of broadcast channel

- **•** $||U|| 1$ is at most number of negative eigenvalues of *H*
- **•** $||V||$ − 1 is at most number of positive eigenvalues of *H*

COMMENTS

It is reasonably straightforward to map this bound on auxiliaries to Marton's inner bound.

The bound $\|W\| \le \|X\| + 4$ comes from convexification (traditional Caratheodory based arguments)

This bound seems to indicate a trade-off between communicating with one receiver vs. the other

• In some sense, the essence of broadcast channel

One can extract more properties of *extremal* $p(u, v, x)$ from the argument: consider the Hessian *H* (of size $||X|| - 1$) of the mapping $p(x) \mapsto H(Y) - H(Z)$

- $||U|| 1$ is at most number of negative eigenvalues of *H*
- $||V|| 1$ is at most number of positive eigenvalues of *H*

- Cover introduced the broadcast channel
- In that he talked about the Spanish-English channel (randomized time-division)
- Later Marton and Cover-van der Meulen came up with improved
- After trying hard and determining the extremal auxiliary random
- May be there is a story behind extremal auxiliary random variables that

- Cover introduced the broadcast channel
- In that he talked about the Spanish-English channel (randomized time-division)
- Later Marton and Cover-van der Meulen came up with improved achievable regions using auxiliary random variables
- After trying hard and determining the extremal auxiliary random
- May be there is a story behind extremal auxiliary random variables that

- Cover introduced the broadcast channel
- In that he talked about the Spanish-English channel (randomized time-division)
- Later Marton and Cover-van der Meulen came up with improved achievable regions using auxiliary random variables
- After trying hard and determining the extremal auxiliary random variables, we see that regions revert to randomized time-division (binary) and an as yet undetermined generalization of randomized time-division.
- May be there is a story behind extremal auxiliary random variables that

- Cover introduced the broadcast channel
- In that he talked about the Spanish-English channel (randomized time-division)
- Later Marton and Cover-van der Meulen came up with improved achievable regions using auxiliary random variables
- After trying hard and determining the extremal auxiliary random variables, we see that regions revert to randomized time-division (binary) and an as yet undetermined generalization of randomized time-division.
- May be there is a story behind extremal auxiliary random variables that will shed light into decades long open problems.

- Cover introduced the broadcast channel
- In that he talked about the Spanish-English channel (randomized time-division)
- Later Marton and Cover-van der Meulen came up with improved achievable regions using auxiliary random variables
- After trying hard and determining the extremal auxiliary random variables, we see that regions revert to randomized time-division (binary) and an as yet undetermined generalization of randomized time-division.
- May be there is a story behind extremal auxiliary random variables that will shed light into decades long open problems.

Thank You