



JACOBS
UNIVERSITY

Outage-free Transmit Power Minimization with Imperfect CSIT

The 2014 Sino-Germany Workshop

"Bridging Theory and Practice in Wireless Communications and Networking"

Giuseppe Abreu

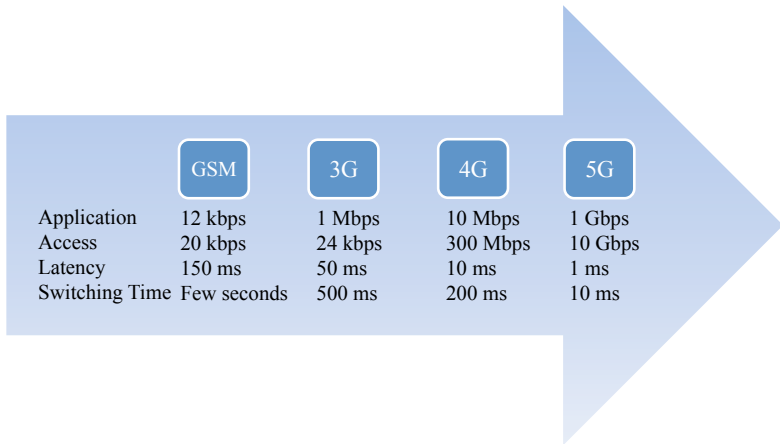
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School of Engineering and Sciences
Jacobs University Bremen

March 4, 2014

Evolution of Cellular Technology

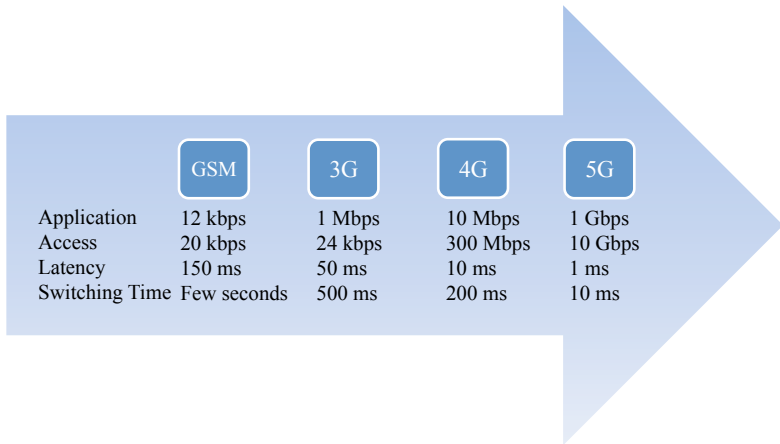
Bridging Theory and Practice in Cellular Systems



	GSM	3G	4G	5G
Application	12 kbps	1 Mbps	10 Mbps	1 Gbps
Access	20 kbps	24 kbps	300 Mbps	10 Gbps
Latency	150 ms	50 ms	10 ms	1 ms
Switching Time	Few seconds	500 ms	200 ms	10 ms

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Trend: 1000× traffic in 10 years!

Past and Future

Bridging Theory and Practice in Cellular Systems

- ▶ **Bridges of old**

Past and Future

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- ▶ **Bridges of old**
 - ▶ 2G - Spectrum

Past and Future

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- ▶ 2G - Spectrum ← “Digital” (GSM/CDMA)

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- ▶ 2G - Spectrum ← “Digital” (GSM/CDMA)
- ▶ 3G - More spectrum

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- ▶ CoMP

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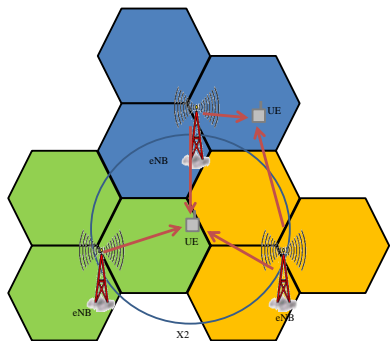
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▶ **Bridge of now** ($4G < ?? < 5G$)

- ▶ CoMP → evolutionary, flexible, mature...

Coordinated Multipoint - CoMP



- ▶ Base stations coordinate with each other.
- ▶ No receiver cooperation.
- ▶ Multiple antennas in transmit and receive side.
- ▶ BSs connected via backhaul and CoMP can be performed:
 - ▶ Joint Processing
 - ▶ Coordinated Beamforming
- ▶ Network Architecture
 - ▶ Centralized Approach
 - ▶ Decentralized Approach

Issues with CoMP

- ▶ The need of tight synchronization between base stations.
- ▶ Signaling overhead on the air interface for the cooperation/coordination of BSs.
- ▶ Backhaul speed and latency for the information exchange between BSs
- ▶ Limitation in the number of cooperating base stations:
Clustering.
- ▶ Sensitivity of the channel information feedback from user terminal to the BSs.

Approaching the problem

- ▶ We consider a multicell multiuser MIMO systems with coordinating BSs.
- ▶ Broader network with conventional size and complexity power.
- ▶ Sufficient resources to estimate the channel.
- ▶ Consider three different problems
 - ▶ Power Minimization Problem **Energy Efficiency**
 - ▶ Max-min SINR Problem **Quality of Service**
 - ▶ Sum Rate Maximization Problem **Spectral Efficiency**

Power Minimization Problem

Energy Efficiency

$$\mathcal{S}_{\text{P1(MISO)}} = \begin{cases} \text{minimize}_{\mathbf{U}, \mathbf{p}} & \sum_{k=1}^K p_k, \\ \text{subject to} & \text{SINR}_k^{\text{DL}} \geq \gamma_k, \quad 1 \leq k \leq K. \end{cases} \quad (1)$$

Max-Min SINR Problem

Quality of Service

$$\mathcal{S}_{P2(\text{MISO})} = \left\{ \begin{array}{l} \underset{\mathbf{U}, \mathbf{p}}{\text{maximize}} \quad \min_{1 \leq k \leq K} \frac{\text{SINR}_k^{\text{DL}}}{\gamma_k} \\ \text{subject to} \quad \sum_k p_k \leq P_{max} \\ \|\mathbf{u}_k\| = 1, \quad 1 \leq k \leq K, \end{array} \right. \quad (2)$$

Sum-Rate Maximization Problem

Spectral Efficiency

$$\mathcal{S}_{\text{P3-MSE(MISO)}} = \begin{cases} \underset{\mathbf{U}, \mathbf{p}}{\text{minimize}} & \sum_{k=1}^K w_k \frac{1}{1 + \text{SINR}^{\text{DL}}} \\ \text{subject to} & \sum_k p_k \leq P_{max} \quad 1 \leq k \leq K. \end{cases} \quad (3)$$

MOTIVATION

Power Minimization Problem

Energy Efficiency

- ▶ Example 1: Power minimization problem [Yu&Lan 2007]

$$\begin{array}{ll} \text{minimize} & \alpha \\ \mathbb{U} \triangleq \{\mathbf{u}_k\} & \end{array}$$

$$\text{subject to } \mathbb{E}|x_k|^2 \leq \alpha p_k$$

$$\text{SINR}_k \geq \gamma_k,$$

$$\text{given } \mathbf{h}_k$$

$$\text{TX : } \mathbf{x}_{N \times 1} = \sum_{k=1}^K s_k \mathbf{u}_k \quad \text{RX : } y_k = \mathbf{h}_k \cdot \mathbf{x} + z_k$$

$$\text{SINR}_k = \frac{|\mathbf{h}_k \cdot \mathbf{u}_k|^2}{\sum_{j \neq k} |\mathbf{h}_k \cdot \mathbf{u}_j|^2 + \sigma^2}$$

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Power Minimization Problem

Energy Efficiency

- ▶ Example 2: Power minimization problem [Song et al. 2007]

$$\underset{\mathbf{p} > 0, \mathbf{V}, \mathbf{U}}{\text{minimize}} \quad \sum_{k=1}^K w_k p_k$$

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- ▶ **Fixed** p_n per-antenna target powers → optimized per user p_k
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Max-min SINR Problem

Quality of Service

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$$\text{SINR}_k = \frac{p_k |\mathbf{h}_k \cdot \mathbf{u}_k|^2}{\sum_{j \neq k} p_j |\mathbf{h}_j \cdot \mathbf{u}_k|^2 + \sigma^2}$$

- ▶ $N = \sum_{b=1}^B N_{t_b}$ TX antennas → MISO
- ▶ ~~Fixed~~ p_n per-antenna target powers → optimized per user p_k
- ▶ ~~Per user~~ γ_k target SINRs → minimum QoS

Max-min SINR Problem

Quality of Service

- ▶ Example 3: Min-max SINR problem [Huang et al. 2011]

$$\text{maximize}_{\mathbf{p} > 0, \mathbb{U}} \quad \min_{\forall k} \text{SINR}_k$$

$$\text{subject to} \quad \|\mathbf{p}\| \leq P$$

$$\|\mathbf{u}_k\| = 1$$

$$\text{given} \quad \mathbf{h}_k$$

$$\text{TX} : \mathbf{x}_{N \times 1} = \sum_{k=1}^K \sqrt{p_k} s_k \mathbf{u}_k \quad \text{RX} : y_k = \mathbf{h}_k \cdot \mathbf{x} + z_k$$

$$\text{SINR}_k = \frac{p_k |\mathbf{h}_k \cdot \mathbf{u}_k|^2}{\sum_{j \neq k} p_j |\mathbf{h}_j \cdot \mathbf{u}_k|^2 + \sigma^2}$$

- ▶ $N = \sum_{b=1}^B N_{t_b}$ TX antennas → MISO
- ▶ ~~Fixed~~ p_n per-antenna target powers → optimized per user p_k
- ▶ ~~Per user~~ γ_k target SINRs → minimum QoS
- ▶ Perfectly known \mathbf{h}_k for all → overhead (!)

Max-min SINR Problem

Quality of Service

- ▶ Example 4: Min-max SINR problem [[Cai et al. 2011](#)]

$$\text{maximize}_{\mathbf{p} > 0, \mathbf{U}, \mathbf{V}} \quad \min_{\forall k} \quad \frac{\text{SINR}_k}{\alpha_k}$$

$$\text{subject to} \quad \mathbf{w}_\ell \cdot \mathbf{p} \leq P_\ell, \ell \in \mathcal{L} \quad \left| \mathcal{L} \right| < K$$

$$\text{given} \quad \mathbf{H}_{kj}, \mathbf{w}_k \text{ and } \alpha$$

$$\text{TX: } \mathbf{x}_{N \times 1} = \sum_{k=1}^K \sqrt{p_k} s_k \mathbf{u}_k \quad \text{RX: } y_k = \mathbf{v}_k^H \cdot \mathbf{H}_{kk} \cdot \mathbf{x} + z_k$$

$$\text{SINR}_k = \frac{p_k |\mathbf{v}_k^H \cdot \mathbf{H}_{kk} \cdot \mathbf{u}_k|^2}{\sum_{j \neq k} p_j |\mathbf{v}_k^H \cdot \mathbf{H}_{kj} \cdot \mathbf{u}_k|^2 + \sigma^2}$$

Max-min SINR Problem

Quality of Service

- ▶ Example 4: Min-max SINR problem [[Cai et al. 2011](#)]

$$\underset{\mathbf{p} > 0, \mathbf{U}, \mathbf{V}}{\text{maximize}} \quad \min_{\forall k} \frac{\text{SINR}_k}{\alpha_k}$$

$$\text{subject to} \quad \mathbf{w}_\ell \cdot \mathbf{p} \leq P_\ell, \ell \in \mathcal{L} \quad \left| \mathcal{L} \right| < K$$

$$\text{given} \quad \mathbf{H}_{kj}, \mathbf{w}_k \text{ and } \alpha$$

$$\text{TX: } \mathbf{x}_{N \times 1} = \sum_{k=1}^K \sqrt{p_k} s_k \mathbf{u}_k \quad \text{RX: } y_k = \mathbf{v}_k^H \cdot \mathbf{H}_{kk} \cdot \mathbf{x} + z_k$$

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- ▶ $N = \sum_{b=1}^B N_{t_b}$ TX antennas \rightarrow MIMO

Max-min SINR Problem

Quality of Service

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$$\text{subject to} \quad \mathbf{w}_\ell \cdot \mathbf{p} \leq P_\ell, \ell \in \mathcal{L} \quad \left| \mathcal{L} \right| < K$$

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- ▶ $N = \sum_{b=1}^B N_{t_b}$ TX antennas → MIMO
- ▶ **Fixed** p_n per-antenna target powers → **optimized per user** p_k
- ▶ **Known** weight vectors per user \mathbf{w}_k and scores α_k → how (?)

Max-min SINR Problem

Quality of Service

- ▶ Example 4: Min-max SINR problem [Cai et al. 2011]

$$\text{maximize}_{\mathbf{p} > 0, \mathbf{U}, \mathbf{V}} \quad \min_{\forall k} \quad \frac{\text{SINR}_k}{\alpha_k}$$

$$\text{subject to} \quad \mathbf{w}_\ell \cdot \mathbf{p} \leq P_\ell, \ell \in \mathcal{L} \quad \left| \mathcal{L} \right| < K$$

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$$\text{TX: } \mathbf{x}_{N \times 1} = \sum_{k=1}^K \sqrt{p_k} s_k \mathbf{u}_k \quad \text{RX: } y_k = \mathbf{v}_k^H \cdot \mathbf{H}_{kk} \cdot \mathbf{x} + z_k$$

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- ▶ $N = \sum_{b=1}^B N_{t_b}$ TX antennas → MIMO
- ▶ ~~Fixed~~ p_n per-antenna target powers → optimized per user p_k
- ▶ ~~Known~~ weight vectors per user \mathbf{w}_k and scores α_k → how (?)
- ▶ ~~Perfectly known~~ \mathbf{H}_{kj} for all users → overhead (!)

Sum-Rate Maximization Problem

Spectral Efficiency

- ▶ Example 5: Sum-rate maximization problem [Tran [et al.](#) 2012]

$$\begin{aligned} & \underset{\mathbf{t}, \mathbb{U}}{\text{maximize}} && \prod_{k=1}^K t_k \\ & \text{subject to} && \text{SINR}_k \geq t_k^{1/\alpha_k} - 1 \\ & && \sum_{k=1}^K \|\mathbf{u}_k\|^2 \leq P \\ & \text{given} && \mathbf{h}_k, P \text{ and } \alpha \end{aligned}$$

$$\alpha_k \log_2(1 + \text{SINR}_k) \longrightarrow (1 + \text{SINR}_k)^{\alpha_k} \longrightarrow t_k$$

Sum-Rate Maximization Problem

Spectral Efficiency

- ▶ Example 5: Sum-rate maximization problem [Tran et al. 2012]

$$\begin{aligned} & \underset{\mathbf{t}, \mathbb{U}}{\text{maximize}} && \prod_{k=1}^K t_k \\ & \text{subject to} && \text{SINR}_k \geq t_k^{1/\alpha_k} - 1 \\ & && \sum_{k=1}^K \|\mathbf{u}_k\|^2 \leq P \\ & \text{given} && \mathbf{h}_k, P \text{ and } \alpha \end{aligned}$$

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- ▶ $N = \sum_{b=1}^B N_{t_b}$ TX antennas \rightarrow MISO

Sum-Rate Maximization Problem

Spectral Efficiency

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- ▶ $N = \sum_{b=1}^B N_{t_b}$ TX antennas \rightarrow MISO
- ▶ **Fixed** p_n per-antenna target powers \rightarrow **optimized total** p_k
- ▶ **Known** scores $\alpha_k \rightarrow$ **how** (?)

Sum-Rate Maximization Problem

Spectral Efficiency

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$$\alpha_k \log_2(1 + \text{SINR}_k) \longrightarrow (1 + \text{SINR}_k)^{\alpha_k} \longrightarrow t_k$$

- ▶ $N = \sum_{b=1}^B N_{t_b}$ TX antennas \rightarrow MISO
- ▶ ~~Fixed p_n per-antenna target powers~~ \rightarrow optimized **total** p_k
- ▶ **Known** scores $\alpha_k \rightarrow$ how (?)
- ▶ **Perfectly known** \mathbf{h}_k for all users \rightarrow overhead (!)

Sum-Rate Maximization Problem

Spectral Efficiency

- ▶ Example 6: Sum-rate maximization problem [Park et al. 2013]

$$\text{maximize}_{\mathbf{V} \triangleq \{\mathbf{V}_k\}} \sum_{k=1}^K w_k \log_2 |\mathbf{I}_{N_r} + (\sigma_k^2 \mathbf{I} + \mathbf{\Phi}_k)^{-1} \mathbf{H}_{kk} \mathbf{V}_k \mathbf{V}_k^H \mathbf{H}_{kk}^H|$$

$$\text{subject to } \|\mathbf{H}_{jk} \mathbf{V}_k\|^2 \leq \alpha_{jk} \sigma_j^2$$

$$\|\mathbf{V}_k\|^2 \leq p_k$$

given \mathbf{H}_{jk} , \mathbf{w} , \mathbf{p} and α

$$\mathbf{\Phi}_k = \sum_{j=1}^K \mathbf{H}_{kj} \mathbf{V}_j \mathbf{V}_j^H \mathbf{H}_{kj}^H$$

Sum-Rate Maximization Problem

Spectral Efficiency

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Sum-Rate Maximization Problem

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$$\mathbf{\Phi}_k = \sum_{j=1}^K \mathbf{H}_{kj} \mathbf{V}_j \mathbf{V}_j^H \mathbf{H}_{kj}^H$$

- ▶ ~~$N = \sum_{b=1}^B N_{t_b}$ TX antennas~~ → MIMO
- ▶ ~~Fixed p_n per-antenna target powers~~ → optimized per user p_k
- ▶ ~~Known weights \mathbf{w} , target powers p_k and scores α_k~~ → how (?)

Sum-Rate Maximization Problem

Spectral Efficiency

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$$\text{subject to } \|\mathbf{H}_{jk} \mathbf{V}_k\|^2 \leq \alpha_{jk} \sigma_j^2$$

$$\|\mathbf{V}_k\|^2 \leq p_k$$

given \mathbf{H}_{jk} , \mathbf{w} , \mathbf{p} and α

$$\mathbf{\Phi}_k = \sum_{j=1}^K \mathbf{H}_{kj} \mathbf{V}_j \mathbf{V}_j^H \mathbf{H}_{kj}^H$$

- ▶ ~~$N = \sum_{b=1}^B N_{t_b}$ TX antennas~~ → MIMO
- ▶ ~~Fixed p_n per-antenna target powers~~ → optimized per user p_k
- ▶ ~~Known weights \mathbf{w} , target powers p_k and scores α_k~~ → how (?)
- ▶ ~~Perfectly known \mathbf{H}_{kj} for all users~~ → overhead (!)

Comprehensive Review

		Instantaneous Perfect CSIT	Instantaneous Imperfect CSIT	Statistical CSIT	Covariance Information
P1	MISO	[20-25, 62]	[26]	[27-30]	[31-33,63]
	MIMO	[22,38,64]			
P2	MISO	[34-37,42]	[43]	[42-44-47]	
	MIMO	[38-41]			
P3	MISO	[48,49,51-55,57,65,66]	[60]	[61]	
	MIMO	[57-59,67-69]			

SOME TOOLS

System Model

- ▶ The transmitted signal

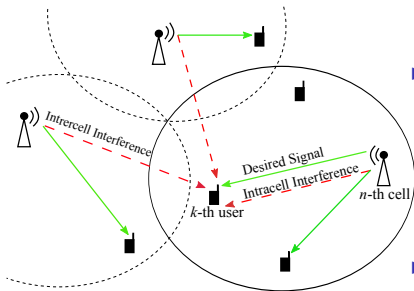
$$\mathbf{x} = \sum_{j=1}^K \sqrt{p_j} \mathbf{u}_j s_j$$

- ▶ The received signal

$$y_k = \mathbf{h}_k^H \left(\sum_{j=1}^K \sqrt{p_j} \mathbf{u}_j s_j \right) + z_k$$

- ▶ The SINR of user k

$$\Phi_k^{\text{DL}} = \frac{p_k |\mathbf{h}_k^H \mathbf{u}_k|^2}{\sum_{j=1, j \neq k}^K p_j |\mathbf{h}_k^H \mathbf{u}_j|^2 + \sigma_{z_k}^2}$$



Perron-Frobenius Theorem

- ▶ Characterizes eigenvectors and eigenvalues of non-negative matrices.
- ▶ Possible solution for power minimization in wireless network.
- ▶ SINR Constraint

$$\frac{p_k |\mathbf{h}_k^H \mathbf{u}_k|^2}{\sum_{j=1, j \neq k}^K p_j |\mathbf{h}_k^H \mathbf{u}_j|^2 + \sigma_{z_k}^2} \geq \gamma_k. \quad (4)$$

- ▶ The SINR constraint can be re-written as

$$p_k G_{kk} \geq \gamma_k \left(\sum_{j=1, j \neq k}^K p_j G_{kj} + \sigma_{z_k}^2 \right), \quad (5)$$

where $G_{jk} = |\mathbf{h}_j \mathbf{u}_k|^2$.

Perron-Frobenius Theorem

- ▶ We now define two matrices \mathbf{D} and \mathbf{G}

$$\mathbf{D} = \begin{bmatrix} \frac{\gamma_1}{G_{11}} & 0 & \cdots & 0 \\ 0 & \frac{\gamma_2}{G_{22}} & \cdots & 0 \\ \vdots & \vdots & \ddots & \vdots \\ 0 & 0 & \cdots & \frac{\gamma_K}{G_{KK}} \end{bmatrix}, \quad \mathbf{G} = \begin{bmatrix} 0 & G_{12} & \cdots & G_{1K} \\ G_{21} & 0 & \cdots & G_{2K} \\ \vdots & \vdots & \ddots & \vdots \\ G_{K1} & G_{K2} & \cdots & 0 \end{bmatrix},$$

- ▶ The SINR constraint in the matrix form can be written as

$$(\mathbf{I} - \mathbf{D}\mathbf{G})\mathbf{p} \geq \mathbf{D}\mathbf{n},$$

where $\mathbf{p} = [p_1, p_2, \dots, p_K]^T$, and $\mathbf{n} = [\sigma_{z_1}^2, \sigma_{z_2}^2, \dots, \sigma_{z_k}^2]^T$.

- ▶ The optimal solution

$$\mathbf{p}^* = (\mathbf{I} - \mathbf{D}\mathbf{G})^{-1}\mathbf{D}\mathbf{n}$$

Perron-Frobenius Theorem

- ▶ The necessary and sufficient conditions for \mathbf{p} to be positive

$$(\mathbf{I} - \underbrace{\mathbf{D}\mathbf{G}}_{\mathbf{A}})^{-1} \geq 0 \quad \text{iff} \quad \rho(\mathbf{A}) = |\lambda_{max}(\mathbf{A})| < 1.$$

Conic Programming

- ▶ Convex optimization tools are widely used in communication / signal processing algorithms.
- ▶ Formulation of the non-convex problems to convex problems.
- ▶ Efficient use of convex optimization tools as CVX and SeDuMi.
- ▶ Consider the SINR constraint

$$p_k |\mathbf{h}_k^H \mathbf{u}_k|^2 - \gamma_k \sum_{j=1, j \neq k}^K p_j |\mathbf{h}_k^H \mathbf{u}_j|^2 \geq \gamma_k (\sigma_{z_k}^2). \quad (6)$$

- ▶ The constraint is a quadratic optimization problem with quadratic non-convex constraints.

Beamforming Problem as SDP

Consider a power minimization problem

$$\mathcal{S}_{\text{P1(MISO)}} = \begin{cases} \text{minimize} & \sum_{k=1}^K \|\mathbf{u}_k\|^2, \\ \text{subject to} & \frac{|\mathbf{h}_k^H \mathbf{u}_k|^2}{\sum_{j=1, j \neq k}^K |\mathbf{h}_k^H \mathbf{u}_j|^2 + \sigma_{z_k}^2} \geq \gamma_k, \quad 1 \leq k \leq K. \end{cases} \quad (7)$$

Let us define $\mathbf{A}_k = \mathbf{u}_k \mathbf{u}_k^H$ and $\mathbf{R}_k = \mathbf{h}_k \mathbf{h}_k^H$. The optimization problem now can be transformed as

$$\mathcal{S}_{\text{P1-SDP(MISO)}} = \begin{cases} \text{minimize} & \sum_{k=1}^K \text{Trace}(\mathbf{A}_k), \\ \text{subject to} & \text{Trace}(\mathbf{R}_k \mathbf{A}_k) - \gamma_k \sum_{j \neq k}^K \text{Trace}(\mathbf{R}_k \mathbf{A}_j) \geq \gamma_k \sigma_{z_k}^2 \\ & \mathbf{A}_k \succeq 0, \quad \text{rank}(\mathbf{A}_k) = 1, \quad 1 \leq k \leq K. \end{cases} \quad (8)$$

Beamforming Problem as SOCP

- ▶ An arbitrary phase rotation does not affect the SINR of the user.
- ▶ Considering only the real part of the gain matrix, the SINR constraint can be re-written as

$$\left(1 + \frac{1}{\gamma_k}\right) |\mathbf{h}_k^H \mathbf{u}_k|^2 \geq \left\| \frac{\mathbf{h}_i^H \mathbf{U}}{\sigma_{z_k}^2} \right\|^2, \quad 1 \leq k \leq K. \quad (9)$$

- ▶ And, the optimization problem thus becomes

$$\mathcal{S}_{\text{P1-SOCP(MISO)}} = \begin{cases} \text{minimize} & \tau \\ \text{subject to} & \sqrt{\left(1 + \frac{1}{\gamma_k}\right)} \mathbf{h}_k^H \mathbf{u}_k \geq \left\| \frac{\mathbf{h}_i^H \mathbf{U}}{\sigma_{z_k}^2} \right\| \\ & \sum_{k=1}^K \|\mathbf{u}_k\| \leq \tau, \quad 1 \leq k \leq K, \end{cases} \quad (10)$$

UL-DL duality via Lagrangian Duality

- ▶ The Lagrangian of the SINR constraint can be written as

$$L(\mathbf{u}_k, \lambda_k) = \sum_{k=1}^K \lambda_k \sigma_{z_k}^2 + \sum_{k=1}^K \mathbf{u}_k^H \left(\mathbf{I} + \sum_{\substack{j=1 \\ j \neq k}}^K \lambda_j \mathbf{h}_j \mathbf{h}_j^H - \frac{\lambda_k}{\gamma_k} \mathbf{h}_k \mathbf{h}_k^H \right) \mathbf{u}_k \quad (11)$$

- ▶ The dual objective is

$$g(\lambda_k) = \min_{\mathbf{u}_k} L(\mathbf{u}_k, \lambda_k) \quad (12)$$

- ▶ The Lagrangian dual problem is

$$\begin{aligned} & \text{maximize} && \sum_{k=1}^K \lambda_k \sigma_{z_k}^2 \\ & \text{subject to} && \sum_{j=1}^K \lambda_j \mathbf{h}_j \mathbf{h}_j^H + \mathbf{I} \succeq \left(1 + \frac{1}{\gamma_k} \right) \lambda_k \mathbf{h}_k \mathbf{h}_k^H \end{aligned} \quad (13)$$

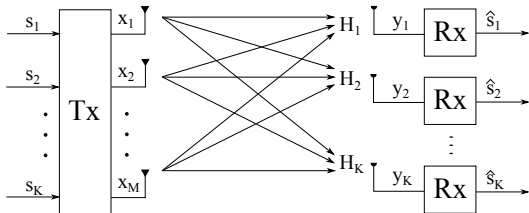
UL-DL duality via Lagrangian Duality

- ▶ The sum power minimization problem for uplink

$$\begin{aligned} & \text{minimize} && \sum_{k=1}^K \rho_k \\ & \text{subject to} && \sum_{j=1}^K \rho_j \mathbf{h}_j \mathbf{h}_j^H + \sigma_{z_k}^2 \mathbf{I} \preceq \left(1 + \frac{1}{\gamma_k}\right) \rho_k \mathbf{h}_k \mathbf{h}_k^H \end{aligned} \tag{14}$$

- ▶ For $\rho_k = \lambda_k \sigma_{z_k}^2$, Eq (13) and Eq (14) are identical.
- ▶ Eq (13) and Eq (14) gives the same solution.
- ▶ The dual variables of the downlink problem have the interpretation of being the uplink power scaled by the noise variance.

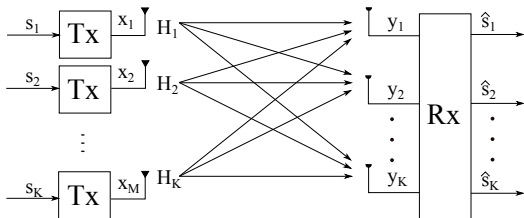
Duality Theory for UL-DL Beamforming



- ▶ The downlink SINR is given as

$$\hat{\Phi}_k^{\text{DL}} = \frac{p_k |\hat{\mathbf{h}}_k^H \mathbf{u}_k|^2}{\sum_{j=1, j \neq k}^K p_j |\hat{\mathbf{h}}_k^H \mathbf{u}_j|^2 + \sigma_{z_i}^2}. \quad (15)$$

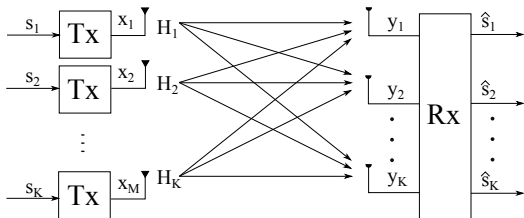
Duality Theory for UL-DL Beamforming



- ▶ The uplink SINR is given as

$$\hat{\Phi}_k^{\text{UL}} = \frac{q_k |\hat{\mathbf{h}}_k^H \mathbf{u}_k|^2}{\mathbf{u}_k^H \left(\sum_{j=1, j \neq k}^K q_j \hat{\mathbf{h}}_j \hat{\mathbf{h}}_j^H + \sigma_{z_k}^2 \mathbf{I} \right) \mathbf{u}_k}. \quad (16)$$

Duality Theory for UL-DL Beamforming



- ▶ The uplink SINR is given as

$$\hat{\Phi}_k^{\text{UL}} = \frac{q_k |\hat{\mathbf{h}}_k^H \mathbf{u}_k|^2}{\mathbf{u}_k^H \left(\sum_{j=1, j \neq k}^K q_j \hat{\mathbf{h}}_j \hat{\mathbf{h}}_j^H + \sigma_{z_k}^2 \mathbf{I} \right) \mathbf{u}_k}. \quad (16)$$

Uplink SINRs are only coupled by transmission powers, however, the downlink SINRs are additionally coupled by beamforming vectors, making direct optimization difficult.

Duality-Downlink Power Assignment

- ▶ For a fixed beamformers, power optimization reduces to

$$\mathcal{S}^{DL}(\tilde{\mathbf{U}}, P_{max}) = \begin{cases} \underset{\mathbf{p}}{\text{maximize}} & \min_{1 \leq k \leq K} \frac{\text{SINR}_k^{(DL)}}{\gamma_k} \\ \text{subject to} & \sum_k p_k = P_{max} \end{cases} \quad (17)$$

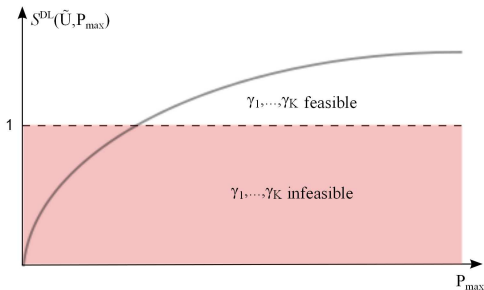
- ▶ $\mathcal{S}^{DL}(\mathbf{U}, P_{max})$ is strictly monotonically increasing in P_{max} .

Duality-Downlink Power Assignment

- ▶ For a fixed beamformers, power optimization reduces to

$$\mathcal{S}^{DL}(\tilde{\mathbf{U}}, P_{max}) = \begin{cases} \text{maximize}_{\mathbf{P}} & \min_{1 \leq k \leq K} \frac{\text{SINR}_k^{(DL)}}{\gamma_k} \\ \text{subject to} & \sum_k p_k = P_{max} \end{cases} \quad (17)$$

- ▶ $\mathcal{S}^{DL}(\mathbf{U}, P_{max})$ is strictly monotonically increasing in P_{max} .



Duality-Downlink Power Assignment

- ▶ If $\tilde{\mathbf{p}}$ is a global maximizer of the optimization problem, then

$$\mathcal{S}^{DL}(\tilde{\mathbf{U}}, P_{max}) = \frac{\text{SINR}_k^{(DL)}(\tilde{\mathbf{U}}, \tilde{\mathbf{p}})}{\gamma_k} \quad (18)$$
$$P_{max} = \|\tilde{\mathbf{p}}\|_1$$

- ▶ Further we can elaborate Eq (18) as

$$\tilde{\mathbf{p}} \frac{1}{\mathcal{S}^{DL}(\tilde{\mathbf{U}}, P_{max})} = \mathbf{D}\mathbf{G}(\tilde{\mathbf{U}})\tilde{\mathbf{p}} + \mathbf{D}\mathbf{n} \quad (19a)$$

$$\frac{1}{\mathcal{S}^{DL}(\tilde{\mathbf{U}}, P_{max})} = \frac{1}{P_{max}} \mathbf{1}^\top \mathbf{D}\mathbf{G}(\tilde{\mathbf{U}})\tilde{\mathbf{p}} + \frac{1}{P_{max}} \mathbf{1}^\top \mathbf{D}\mathbf{n} \quad (19b)$$

- ▶ We can form eigen-system as

$$\underbrace{\begin{bmatrix} \mathbf{D}\mathbf{G}(\tilde{\mathbf{U}}) & \mathbf{D}\mathbf{n} \\ \frac{1}{P_{max}} \mathbf{1}^\top \mathbf{D}\mathbf{G}(\tilde{\mathbf{U}}) & \frac{1}{P_{max}} \mathbf{1}^\top \mathbf{D}\mathbf{n} \end{bmatrix}}_{\Upsilon(\tilde{\mathbf{U}}, P_{max})} \begin{bmatrix} \tilde{\mathbf{p}} \\ 1 \end{bmatrix} = \frac{1}{\mathcal{S}^{DL}(\tilde{\mathbf{U}}, P_{max})} \begin{bmatrix} \tilde{\mathbf{p}} \\ 1 \end{bmatrix} \quad (20)$$

Duality-Downlink Power Assignment

- ▶ The solution for SINR balancing problem (17) is now given as

$$\mathcal{S}^{DL}(\tilde{\mathbf{U}}, P_{max}) = \frac{1}{\lambda_{\max}(\Upsilon(\tilde{\mathbf{U}}, P_{max}))} \quad (21)$$

- ▶ Similarly for uplink, and $\tilde{\mathbf{q}}_{\text{ext}} = \begin{pmatrix} \tilde{\mathbf{q}} \\ 1 \end{pmatrix}$

$$\underbrace{\begin{bmatrix} \mathbf{D}\mathbf{G}^T(\tilde{\mathbf{U}}) & \mathbf{D}\mathbf{n} \\ \frac{1}{P_{max}}\mathbf{1}^T\mathbf{D}\mathbf{G}^T(\tilde{\mathbf{U}}) & \frac{1}{P_{max}}\mathbf{1}^T\mathbf{D}\mathbf{n} \end{bmatrix}}_{\Lambda(\tilde{\mathbf{U}}, P_{max})} \tilde{\mathbf{q}}_{\text{ext}} = \lambda_{\max}(\Lambda(\tilde{\mathbf{U}}, P_{max})) \tilde{\mathbf{q}}_{\text{ext}} \quad (22)$$

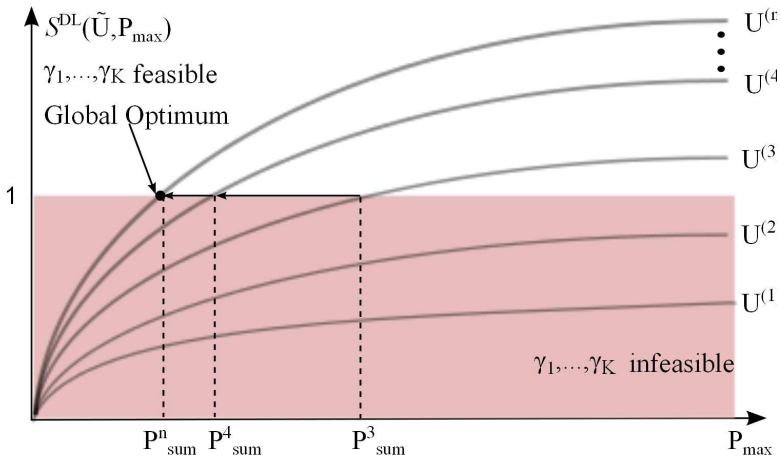
Duality-SINR Balancing

- ▶ For a given \mathbf{q}_{ext} , the cost function $\lambda(\mathbf{U}, \mathbf{q}_{\text{ext}})$ is minimized by independent maximization of the uplink SINRs.
- ▶ The optimal \mathbf{u}_k could be now calculated as

$$\hat{\mathbf{u}}_k = \arg \max_{\mathbf{u}_k} \frac{\mathbf{u}_k^H \overbrace{\mathbf{h}_k \mathbf{h}_k^H}^{\mathbf{H}_k} \mathbf{u}_k}{\underbrace{\mathbf{u}_k^H \left(\sum_{j=1, j \neq k}^K q_j \hat{\mathbf{h}}_j \hat{\mathbf{h}}_j^H + \sigma_{z_k}^2 \mathbf{I} \right) \mathbf{u}_k}_{\mathbf{W}_k}}. \quad (23)$$

- ▶ Eq. (23) is maximizing Rayleigh quotient problem.
- ▶ Eq. (23) is solved via dominant generalized eigen-vectors of matrix pairs $(\mathbf{H}_k, \mathbf{W}_k)$.

Power Minimization Algorithm



WHAT IF CHANNEL IS NOT PERFECT?

System Model

- ▶ MISO channel setting with N_t transmit antennas and K users with single antenna.
- ▶ Consider the transmit beamforming vector \mathbf{u}_k .
- ▶ Transmit signal from BS

$$\mathbf{x} = \sum_{k=1}^K \mathbf{u}_k s_k, \quad (24)$$

- ▶ Received signal at user k

$$y_k = \mathbf{h}_k^H \left(\sum_{j=1}^K s_j \mathbf{u}_j \right) + z_k, \quad (25)$$

- ▶ The instantaneous SINR for user k

$$\text{SINR}_k(\Phi_k) = \frac{|\mathbf{h}_k^H \mathbf{u}_k|^2}{\sum_{j \neq k} |\mathbf{h}_k^H \mathbf{u}_j|^2 + \sigma_{z_k}^2}. \quad (26)$$

Channel Error

- ▶ Common assumption:

Channel Error

- ▶ Common assumption:
BS has perfect knowledge of channel

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- ▶ Reality:

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No perfect CSIT available and constitute error within.

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Estimation errors and feedback delays.

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- ▶ What can be done?

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- ▶ Reality:
No perfect CSIT available and constitute error within.
- ▶ Causes:
Estimation errors and feedback delays.
- ▶ Effects:
BSs cannot predict exactly the required SINR at the users.
- ▶ What can be done?
Estimate SINR under imperfect CSIT

Channel Error

- ▶ We model the imperfect CSIT as

$$\mathbf{h}_k = \hat{\mathbf{h}}_k + \mathbf{e}_{\mathbf{h}_k}, \quad (27)$$

$\hat{\mathbf{h}}_k$ is the estimated channel, $\mathbf{e}_{\mathbf{h}_k}$ is the respective channel error.

- ▶ The SINR with channel error incorporated will be

$$\Phi_{k(\text{error})} = \frac{|(\hat{\mathbf{h}}_k + \mathbf{e}_{\mathbf{h}_k})^H \mathbf{u}_k|^2}{\sum_{j \neq k} |(\hat{\mathbf{h}}_k + \mathbf{e}_{\mathbf{h}_k})^H \mathbf{u}_j|^2 + \sigma_{z_k}^2}. \quad (28)$$

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- ▶ **The problem again?** We still do not know the errors.

SINR Estimate under imperfect CSIT

- ▶ The estimated received signal

$$\begin{aligned}\hat{y}_k &= \hat{\mathbf{h}}_k^H \mathbf{u}_k s_k + \mathbf{e}_{h_k}^H \mathbf{u}_k s_k + \underbrace{\sum_{j=1, j \neq k}^K \hat{\mathbf{h}}_k^H \mathbf{u}_j s_j}_{\text{estimated interference}} + \sum_{j=1, j \neq k}^K \mathbf{e}_{h_k}^H s_j \mathbf{u}_j + z_k, \\ &= \underbrace{\hat{\mathbf{h}}_k^H \mathbf{u}_k s_k}_{\text{estimated transmit signal}} + \sum_{j=1, j \neq k}^K \hat{\mathbf{h}}_k^H \mathbf{u}_j s_j + \underbrace{\sum_{k=1,}^K \mathbf{e}_{h_k}^H s_k \mathbf{u}_k}_{\text{unknown interference}} + z_k.\end{aligned}$$

SINR Estimate under imperfect CSIT

- ▶ The estimated received signal

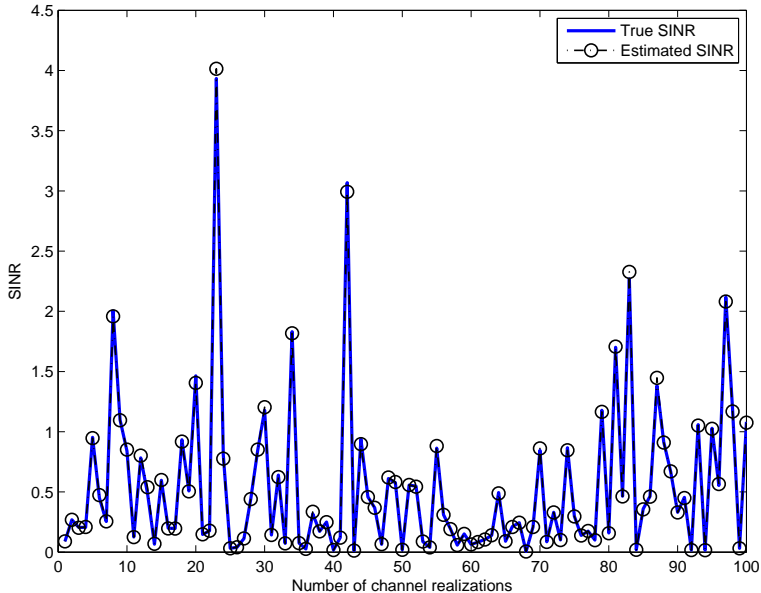
$$\begin{aligned}
 \hat{y}_k &= \hat{\mathbf{h}}_k^H \mathbf{u}_k s_k + \mathbf{e}_{h_k}^H \mathbf{u}_k s_k + \underbrace{\sum_{j=1, j \neq k}^K \hat{\mathbf{h}}_k^H \mathbf{u}_j s_j}_{\text{estimated interference}} + \sum_{j=1, j \neq k}^K \mathbf{e}_{h_k}^H s_j \mathbf{u}_j + z_k, \\
 &= \underbrace{\hat{\mathbf{h}}_k^H \mathbf{u}_k s_k}_{\text{estimated transmit signal}} + \sum_{j=1, j \neq k}^K \hat{\mathbf{h}}_k^H \mathbf{u}_j s_j + \underbrace{\sum_{k=1,}^K \mathbf{e}_{h_k}^H s_k \mathbf{u}_k}_{\text{unknown interference}} + z_k.
 \end{aligned}$$

- ▶ The biased estimated of SINR will be now

$$\hat{\Phi}_{k(\text{biased})} = \frac{|\hat{\mathbf{h}}_k^H \mathbf{u}_k|^2}{\sum_{j=1, j \neq k}^K |\hat{\mathbf{h}}_k^H \mathbf{u}_j|^2 + \sigma_{\mathbf{e}_{h_k}}^2 \text{Trace}(\mathbf{U}\mathbf{U}^H) + \sigma_{z_k}^2} \quad (30)$$

Robustness of Estimate

SINR for different channel realizations for $N_t = 4, K = 4, \sigma_e^2 = 0.01$



Comparison against unbiased

- ▶ The unbiased estimation of SINR

$$\hat{\Phi}_{k(\text{unbiased})} = \frac{|(\hat{\mathbf{h}}_k)^H \mathbf{u}_k|^2}{\sum_{j \neq k} |(\hat{\mathbf{h}}_k)^H \mathbf{u}_j|^2 + \sigma_{z_k}^2}. \quad (31)$$

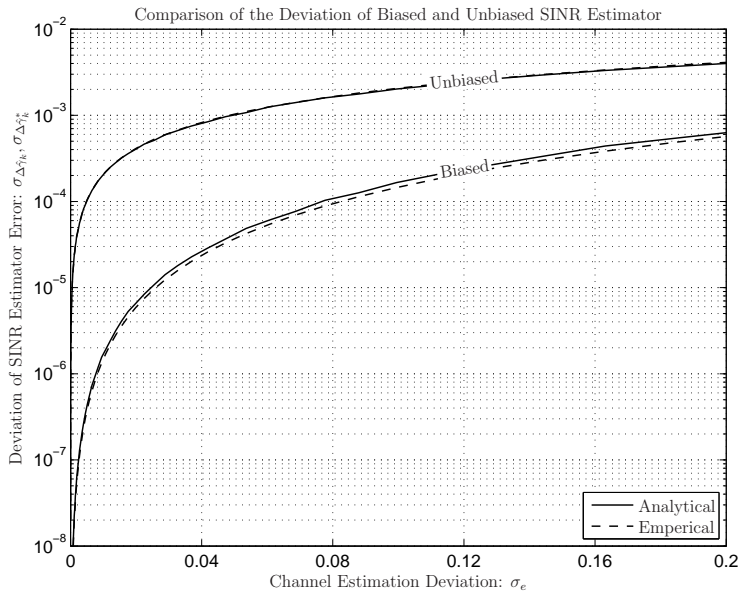
- ▶ Consider estimation errors of unbiased and biased estimations.

$$\begin{aligned} \Delta \hat{\Phi}_{k(\text{unbiased})} &\triangleq \hat{\Phi}_{k(\text{unbiased})} - \Phi_k, \\ \Delta \hat{\Phi}_{k(\text{biased})} &\triangleq \hat{\Phi}_{k(\text{biased})} - \Phi_k, \end{aligned}$$

- ▶ The deviation of error holds following equality

$$\sigma_{\Delta \hat{\Phi}_{k(\text{biased})}} = \frac{\sigma_{\mathbf{e}_{\mathbf{h}_k}}}{\sqrt{2} \left((1 + \sigma_{\mathbf{e}_{\mathbf{h}_k}}^2) + \frac{(\sigma_{z_k}^2 - \text{Trace}(\mathbf{u}_k \mathbf{u}_k^H))}{\text{Trace}(\mathbf{U} \mathbf{U}^H)} \right)} \sigma_{\Delta \hat{\Phi}_{k(\text{unbiased})}}. \quad (33)$$

Comparison against unbiased



Back to Power Minimization Problem

- ▶ The SINR constraint in presence of channel error is

$$p_k G_{kk} \geq \gamma_k \left(\sum_{j=1, j \neq k}^K p_j G_{kj} + \sum_k p_k \sigma_{\mathbf{e}_{\mathbf{h}_k}}^2 + \sigma_{z_k}^2 \right). \quad (34)$$

where $G_{kj} = |\hat{\mathbf{h}}_k^H \mathbf{u}_j|^2$.

- ▶ We now define for $a_k = \sigma_{\mathbf{e}_{\mathbf{h}_k}}^2$

$$\mathbf{D} = \begin{bmatrix} \frac{\gamma_1}{G_{11}} & 0 & \cdots & 0 \\ 0 & \frac{\gamma_2}{G_{22}} & \cdots & 0 \\ \vdots & \vdots & \ddots & \vdots \\ 0 & 0 & \cdots & \frac{\gamma_K}{G_{KK}} \end{bmatrix}, \quad \mathbf{G}_\mathbf{I} = \begin{bmatrix} a_1 & (G_{12} + a_2) & \cdots & G_{1K} + a_K \\ (G_{21} + a_1) & a_2 & \cdots & G_{2K} + a_K \\ \vdots & \vdots & \ddots & \vdots \\ (G_{K1} + a_1) & (G_{K2} + a_2) & \cdots & a_K \end{bmatrix}. \quad (35)$$

- ▶ In matrix form, the SINR constraint can be written as

$$(\mathbf{I} - \mathbf{D}\mathbf{G}_\mathbf{I})\mathbf{p} \geq \mathbf{D}\mathbf{n}, \quad (36)$$

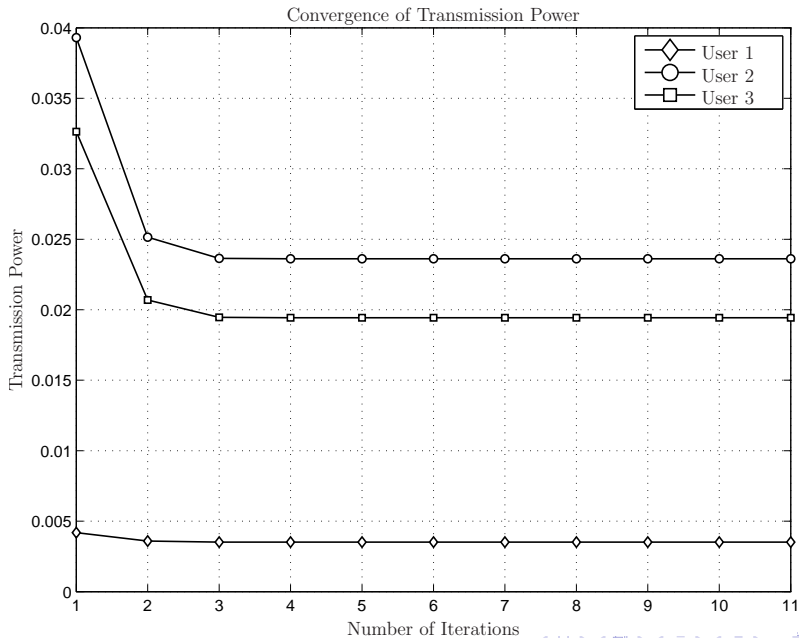
Power Minimization Problem

- ▶ The optimal \mathbf{u}_k can be calculated as 'maximizing Rayleigh quotient' problem

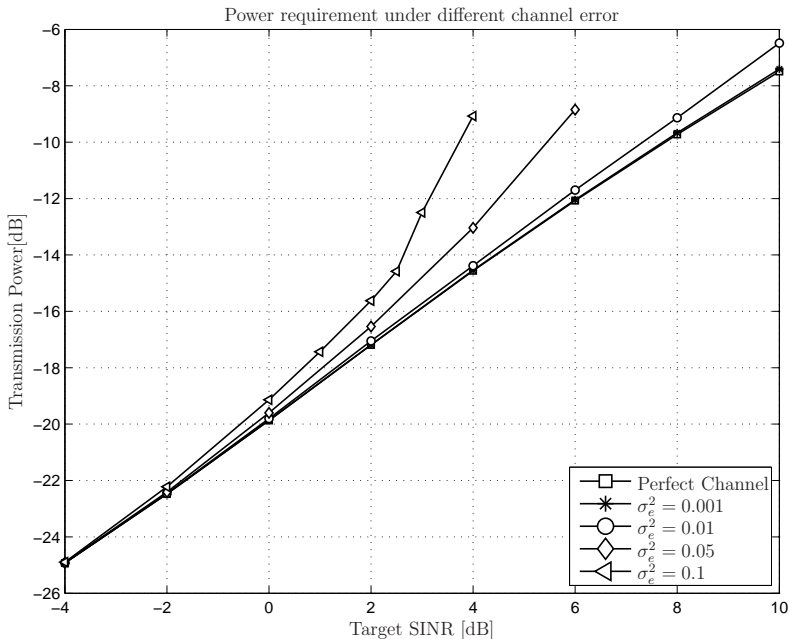
$$\hat{\mathbf{u}}_k = \arg \max_{\mathbf{u}_k} \frac{\mathbf{u}_k^H \underbrace{\hat{\mathbf{h}}_k \hat{\mathbf{h}}_k^H}_{\mathbf{H}_k} \mathbf{u}_k}{\mathbf{u}_k^H \underbrace{\left(\sum_{j=1, j \neq k}^K q_j \hat{\mathbf{h}}_j \hat{\mathbf{h}}_j^H + \sigma_{z_k}^2 \mathbf{I} + \sum_k q_k \sigma_{e_{\mathbf{h}_k}}^2 \mathbf{I} \right)}_{\mathbf{W}_k} \mathbf{u}_k}. \quad (37)$$

- ▶ The optimal \mathbf{u} 's are given by the dominant generalized eigenvectors of matrix pairs $(\mathbf{H}_k, \mathbf{W}_k)$.

Results: Convergence of Algorithm



Results: Comparison of different errors



Future Works

- ▶ The stochastic SINR (for high SNR) for channel error is given as

$$\hat{\Phi}_{k(\text{biased})} = \frac{|\hat{\mathbf{h}}_k^H \mathbf{u}_k|^2}{\sum_{j=1, j \neq k}^K |\hat{\mathbf{h}}_k^H \mathbf{u}_j|^2 + \underbrace{\mathbf{e}_{h_k}^H \mathbf{U} \mathbf{U}^H \mathbf{e}_{h_k}}_{X_1} + \sigma_{z_k}^2} \quad (38)$$

- ▶ X_1 is gamma distributed $\Gamma(x_1; a, b)$ with shape parameter $a \geq 0$, and scale parameter $b \geq 0$.
- ▶ The density function is given as

$$p_{X_1}(x_1; a, b) = \frac{1}{b^a \Gamma(a)} x_1^{a-1} e^{\left(\frac{-x_1}{b}\right)}. \quad (39)$$

- ▶ The parameters a and b is given as

$$a = \frac{(\text{Tr}[\mathbf{U} \mathbf{U}^H])^2}{\text{Tr}[(\mathbf{U} \mathbf{U}^H)^2]} \quad (40a)$$

$$b = \frac{\sigma_{e_{h_k}}^2 \text{Tr}[(\mathbf{U} \mathbf{U}^H)^2]}{\text{Tr}[\mathbf{U} \mathbf{U}^H]} \quad (40b)$$

Future Works-Estimating Distribution

- ▶ Based on distribution of X_1 , the estimated SINR is distributed as

$$p_{\hat{\Phi}_k}(\hat{\phi}, a, b) = \frac{c}{\hat{\phi}^2} \frac{\left(\frac{c}{\hat{\phi}} - \delta\right)^{a-1} \exp\left(-\frac{c/\hat{\phi} - \delta}{b}\right)}{b^a \Gamma(a)}, \quad (41)$$

where the constants c and δ are given as

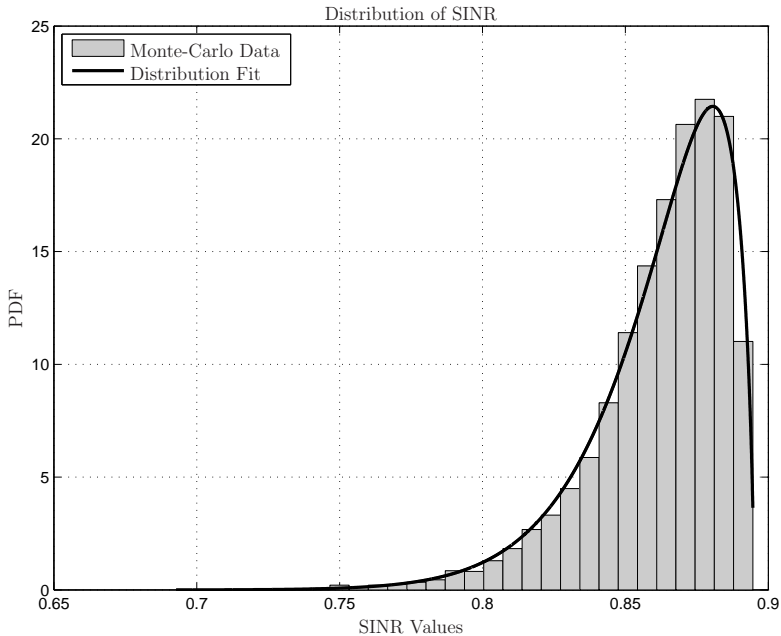
$$c = |\hat{\mathbf{h}}_k^H \mathbf{u}_k|^2, \quad (42a)$$

$$\delta = \sum_{j=1, j \neq k}^K |\hat{\mathbf{h}}_k^H \mathbf{u}_j|^2 + \sigma_{z_k}^2, \quad (42b)$$

- ▶ As an example, the probabilistic constrained power minimization problem can be written as

$$\mathcal{S}_{\text{Prob(MISO)}} = \begin{cases} \underset{\mathbf{P}}{\text{minimize}} & \sum_{k=1}^K \|\mathbf{u}_k\|^2 \\ \text{subject to} & \text{Prob}\{\text{SINR}_k \geq \gamma_k\} \geq 1 - \rho_k \quad 1 \leq k \leq K, \end{cases} \quad (43)$$

Estimating Distribution of SINR



Future Works

- ▶ The channel error case could be now extended to probabilistic approach.
- ▶ Consider different channel errors: uncertainty region bounded / not bounded.
- ▶ Extend the 'Power Minimization', 'Max-min SINR', 'Sum-Rate Maximization' problem to MIMO multi-cell cases.
- ▶ Implement 'Centralized' and 'Decentralized' processing schemes.
- ▶ Compare all three problems under same umbrella.

Thank You

Questions?

Suggestions!